

1A Ultra Low Dropout Linear Regulator with Programmable Current Limiting

ISL80121-5

The ISL80121-5 is a low dropout voltage, single output LDO with programmable current limiting. This LDO operates from input voltages of 2.2V to 6V. The ISL80121-5 has a nominal output voltage of 5V. Other custom voltage options are available upon request.

A sub-micron BiCMOS process is utilized for this product family to deliver the best in class analog performance and overall value. The programmable current limiting improves system reliability of end applications. An external capacitor on the soft-start pin provides an adjustable soft-starting ramp. The ENABLE feature allows the part to be placed into a low quiescent current shutdown mode.

This CMOS LDO will consume significantly lower quiescent current as a function of load compared to bipolar LDOs, which translates into higher efficiency and packages with smaller footprints. Quiescent current is modestly compromised to achieve a very fast load transient response.

TABLE 1. KEY DIFFERENCES BETWEEN FAMILY OF PARTS

PART NUMBER	I _{LIMIT} (DEFAULT)	PROGRAMMABLE I _{SET}
ISL80101	1.75A	No
ISL80101A	1.62A	Yes
ISL80121-5	0.75A	Yes

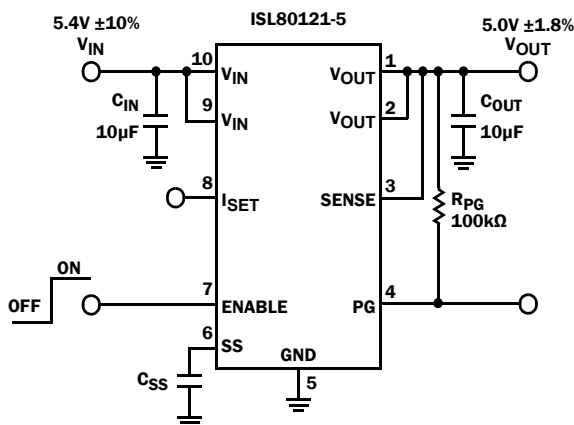
Features

- $\pm 1.8\%$ V_{OUT} Accuracy Guaranteed Over Line, Load and T_J = -40°C to +125°C
- Very Low 130mV Dropout Voltage at V_{IN} = 5.0V
- High Accuracy Current Limit Programmable up to 1.75A
- Very Fast Transient Response
- 210μV_{RMS} Output Noise
- Power-Good Output
- Programmable Soft-Start
- Over-Temperature Protection
- Small 10 Ld DFN Package

Applications

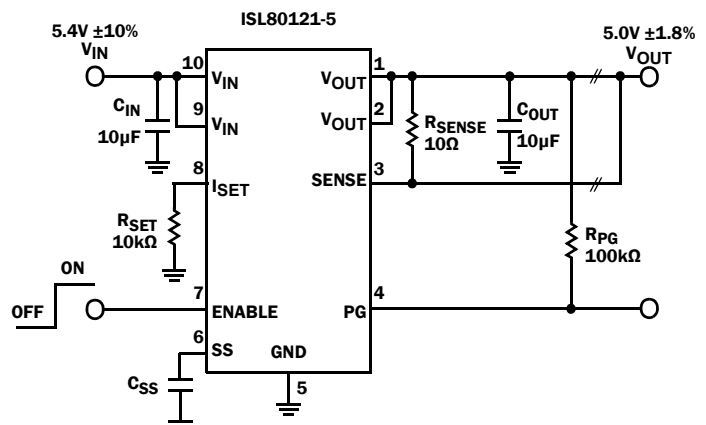
- USB devices
- Telecommunications and Networking
- Medical Equipment
- Instrumentation Systems
- Routers and Switchers
- Gaming

Typical Applications



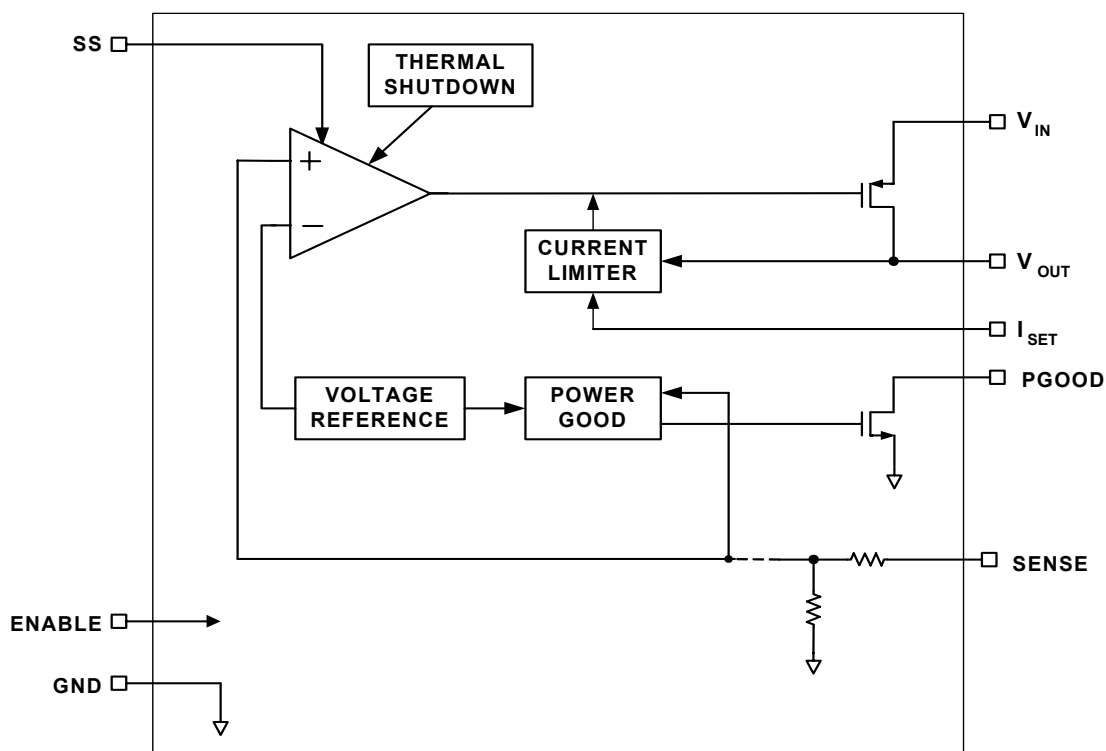
$$I_{LIMIT} = 0.75A$$

(default)



$$I_{LIMIT} = 0.75 + \frac{2.9}{R_{SET}(k\Omega)}$$

Block Diagram



Ordering Information

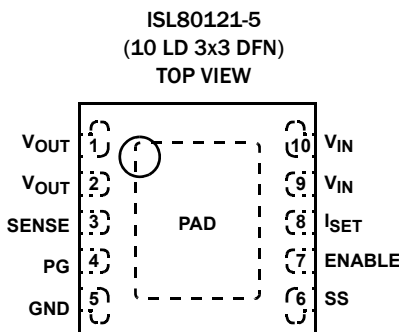
PART NUMBER (Notes 1, 2, 4)	PART MARKING	V _{OUT} VOLTAGE (Note 3)	TEMP. RANGE (°C)	PACKAGE (Pb-Free)	PKG DWG. #
ISL80121IR50Z	DZAD	5.0V	-40 to +125	10 Ld 3x3 DFN	L10.3x3

NOTES:

1. Add "-T*" suffix for tape and reel. Please refer to [TB347](#) for details on reel specifications.
2. These Intersil Pb-free plastic packaged products employ special Pb-free material sets, molding compounds/die attach materials, and 100% matte tin plate plus anneal (e3 termination finish, which is RoHS compliant and compatible with both SnPb and Pb-free soldering operations). Intersil Pb-free products are MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.
3. The 1.5V, 3.3V and 5V fixed output voltages will be released in the future. Please contact Intersil Marketing for more details.
4. For Moisture Sensitivity Level (MSL), please see device information page for [ISL80121-5](#). For more information on MSL, please see Technical Brief [TB363](#).

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Pin Configuration



Pin Descriptions

PIN NUMBER	PIN NAME	DESCRIPTION
1, 2	V _{OUT}	Output voltage. A minimum 10 μ F X5R/X7R output capacitor is required for stability. See “External Capacitor Requirements” on page 8 in the “Functional Description” for more details.
3	SENSE	Remote voltage sense for internally fixed V _{OUT} options. Parasitic resistance between the V _{OUT} pin and the load causes small voltage drops which degrade V _{OUT} accuracy. For applications that require a stiff V _{OUT} , connect the sense pin to the load.
4	PG	V _{OUT} in regulation signal. Logic low indicates V _{OUT} is not in regulation, and must be grounded if not used.
5	GND	Ground.
6	SS	External capacitor adjusts in-rush current.
7	ENABLE	V _{IN} -independent chip enable. TTL and CMOS compatible.
8	I _{SET}	Current limit setting. Current limit is 0.75A when this pin is left floating. This default value can be increased by tying R _{SET} to GND, or decreased by tying R _{SET} to V _{IN} . See “Programmable Current Limit” on page 7 in the “Functional Description” for more details. Do not short this pin to ground.
9, 10	V _{IN}	Input supply. A minimum of 10 μ F X5R/X7R input capacitor is required for stability. See “External Capacitor Requirements” on page 8 in “Functional Description” for more details.
-	EPAD	EPAD at ground potential. Soldering it directly to GND plane is required for thermal considerations. See “Power Dissipation and Thermals” on page 9 for more details.

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Absolute Maximum Ratings (Note 7)

V_{IN} Relative to GND	-0.3V to +6.5V
V_{OUT} Relative to GND	-0.3V to +6.5V
PG, ENABLE, SENSE, SS, I_{SET}	
Relative to GND	-0.3V to +6.5V
ESD Rating	
Human Body Model (Tested per JESD22-A114)	2.5kV
Machine Model (Tested per JESD22-A115)	250V
Latch Up (Tested per JESD78)	$\pm 100\text{mA}$ @ 85°C

Thermal Information

Thermal Resistance (Typical)	θ_{JA} (°C/W)	θ_{JC} (°C/W)
10 Ld 3x3 DFN Package (Notes 5, 6)	48	7
Maximum Junction Temperature (Plastic Package)	+150°C	
Storage Temperature Range	-65°C to +150°C	
Pb-Free Reflow Profile	see link below	
	http://www.intersil.com/pbfree/Pb-FreeReflow.asp	

Recommended Operating Conditions (Note 8)

Junction Temperature Range (T_J)	-40°C to +125°C
V_{IN} Relative to GND	2.2V to 6V
V_{OUT} Range	800mV to 5V
PG, ENABLE, SENSE, SS, I_{SET} Relative to GND	0V to 6V
PG Sink Current	10mA

CAUTION: Do not operate at or near the maximum ratings listed for extended periods of time. Exposure to such conditions may adversely impact product reliability and result in failures not covered by warranty.

NOTES:

- θ_{JA} is measured in free air with the component mounted on a high effective thermal conductivity test board with “direct attach” features. See Tech Brief TB379.
- For θ_{JC} , the “case temp” location is the center of the exposed metal pad on the package underside.
- Absolute maximum voltage rating is defined as the voltage applied for a lifetime average duty cycle above 6V of 1%.
- Electromigration specification defined as lifetime average junction temperature of +110°C where max rated DC current = lifetime average current.

Electrical Specifications

Unless otherwise noted, all parameters are established over the following specified conditions: $V_{IN} = V_{OUT} + 0.4\text{V}$, $V_{OUT} = 5.0\text{V}$, $C_{IN} = C_{OUT} = 10\mu\text{F}$, $T_J = +25^\circ\text{C}$, $I_{LOAD} = 0\text{A}$. Applications must follow thermal guidelines of the package to determine worst case junction temperature. Please refer to “Functional Description” on page 7 and Tech Brief [TB379](#).

Boldface limits apply over the operating temperature range, -40°C to +125°C. Pulse load techniques used by ATE to ensure $T_J = T_A$ defines established limits.

PARAMETER	SYMBOL	TEST CONDITIONS	MIN (Note 9)	TYP	MAX (Note 9)	UNITS
DC CHARACTERISTICS						
DC Output Voltage Accuracy	V_{OUT}	$V_{OUT} + 0.4\text{V} < V_{IN} < 6\text{V}$; $0\text{A} < I_{LOAD} < 1\text{A}$	-1.8		1.8	%
DC Input Line Regulation	$\Delta V_{OUT}/\Delta V_{IN}$	$V_{OUT} + 0.4\text{V} < V_{IN} < 6.0\text{V}$, $V_{OUT} = 5.0\text{V}$			1	%
DC Output Load Regulation	ΔV_{OUT}	$0\text{A} < I_{LOAD} < 1\text{A}$	-1			%
Ground Pin Current	I_Q	$I_{LOAD} = 0\text{A}$, $2.2\text{V} < V_{IN} < 6\text{V}$		3	5	mA
		$I_{LOAD} = 1\text{A}$, $2.2\text{V} < V_{IN} < 6\text{V}$		5	7	mA
Ground Pin Current in Shutdown	I_{SHDN}	ENABLE = 0.2V, $V_{IN} = 6\text{V}$		0.2	12	μA
Dropout Voltage (Note 10)	V_{DO}	$I_{LOAD} = 1\text{A}$, $V_{IN} = 5.0\text{V}$, $V_{SENSE} = 0\text{V}$		90	130	mV
Output Current Limit	I_{LIMIT}	$V_{OUT} = 4.75\text{V}$, $V_{OUT} + 0.4\text{V} < V_{IN} < 6\text{V}$, I_{SET} is floating	0.66	0.75	0.84	A
		$V_{OUT} = 4.75\text{V}$, $V_{OUT} + 0.4\text{V} < V_{IN} < 6\text{V}$, $R_{SET} = 19.33\text{k}\Omega$		0.9		A
Thermal Shutdown Temperature	TSD	$V_{OUT} + 0.4\text{V} < V_{IN} < 6\text{V}$		160		°C
Thermal Shutdown Hysteresis (Rising Threshold)	TSDn	$V_{OUT} + 0.4\text{V} < V_{IN} < 6\text{V}$		30		°C
AC CHARACTERISTICS						
Input Supply Ripple Rejection	PSRR	$f = 1\text{kHz}$, $I_{LOAD} = 1\text{A}$		40		dB
		$f = 1\text{kHz}$, $I_{LOAD} = 100\text{mA}$		40		dB
Output Noise Voltage		$I_{LOAD} = 10\text{mA}$, $\text{BW} = 10\text{Hz} < f < 100\text{kHz}$		210		μV_{RMS}
ENABLE PIN CHARACTERISTICS						
Turn-on Threshold	$V_{EN(HIGH)}$	$2.2\text{V} < V_{IN} < 6\text{V}$	0.3	0.8	1.0	V
Hysteresis (Rising Threshold)	$V_{EN(HYS)}$	$2.2\text{V} < V_{IN} < 6\text{V}$	10	80	200	mV

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Electrical Specifications

Unless otherwise noted, all parameters are established over the following specified conditions: $V_{IN} = V_{OUT} + 0.4V$, $V_{OUT} = 5.0V$, $C_{IN} = C_{OUT} = 10\mu F$, $T_J = +25^\circ C$, $I_{LOAD} = 0A$. Applications must follow thermal guidelines of the package to determine worst case junction temperature. Please refer to "Functional Description" on page 7 and Tech Brief [TB379](#).

Boldface limits apply over the operating temperature range, $-40^\circ C$ to $+125^\circ C$. Pulse load techniques used by ATE to ensure $T_J = T_A$ defines established limits. (Continued)

PARAMETER	SYMBOL	TEST CONDITIONS	MIN (Note 9)	TYP	MAX (Note 9)	UNITS
ENABLE Pin Turn-on Delay	t_{EN}	$C_{OUT} = 10\mu F$, $I_{LOAD} = 1A$		100		μs
ENABLE Pin Leakage Current		$V_{IN} = 6V$, $ENABLE = 3V$			1	μA
SOFT-START CHARACTERISTICS						
Reset Pull-Down Current	I_{PD}	$ENABLE = 0V$, $SS = 1V$	0.5	1	1.3	mA
Soft-Start Charge Current	I_{CHG}		-3.3	-2	-0.8	μA
PG PIN CHARACTERISTICS						
V_{OUT} PG Flag Threshold			75	84	92	$\%V_{OUT}$
V_{OUT} PG Flag Hysteresis				4		%
PG Flag Low Voltage		$I_{SINK} = 500\mu A$		47	100	mV
PG Flag Leakage Current		$V_{IN} = 6V$, $PG = 6V$		0.05	1	μA

NOTES:

- Compliance to datasheet limits is assured by one or more methods: production test, characterization and/or design.
- Dropout is defined by the difference in supply V_{IN} and V_{OUT} when the output is below its nominal regulation.

Typical Operating Performance

Unless otherwise noted: $V_{IN} = 5.4V$, $V_{OUT} = 5.0V$, $C_{IN} = C_{OUT} = 10\mu F$, $T_J = +25^\circ C$, $I_L = 0A$.

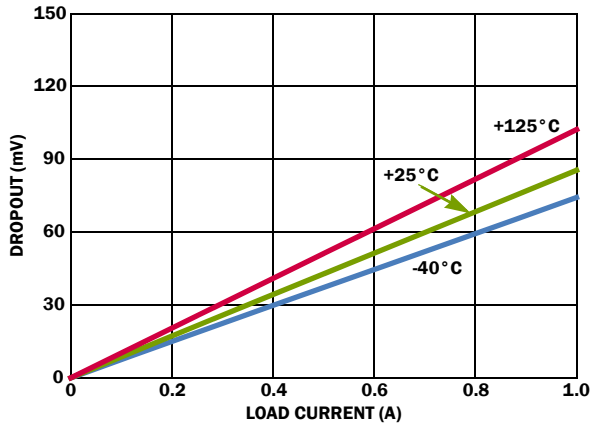


FIGURE 1. DROPOUT VOLTAGE vs LOAD

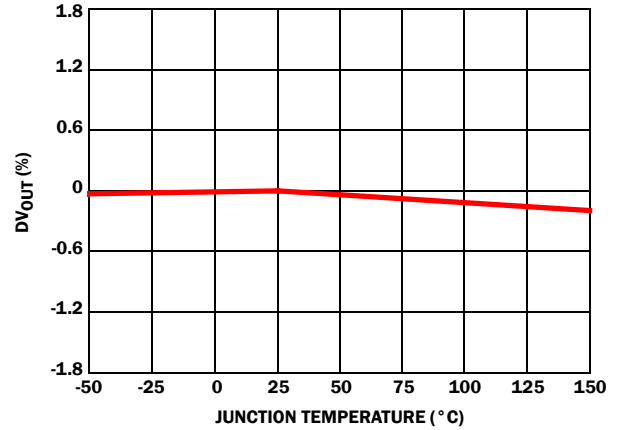


FIGURE 2. OUTPUT VOLTAGE vs TEMPERATURE

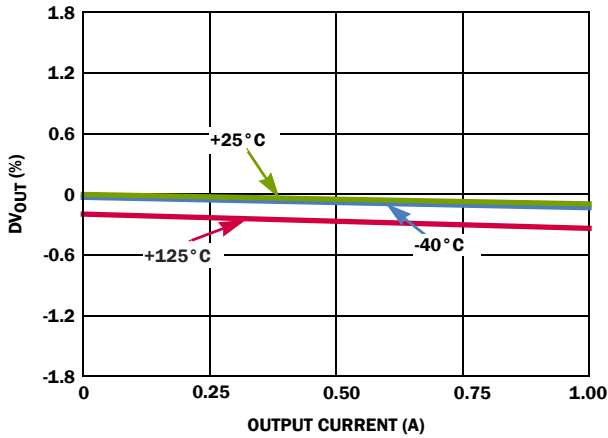


FIGURE 3. OUTPUT VOLTAGE vs OUTPUT CURRENT

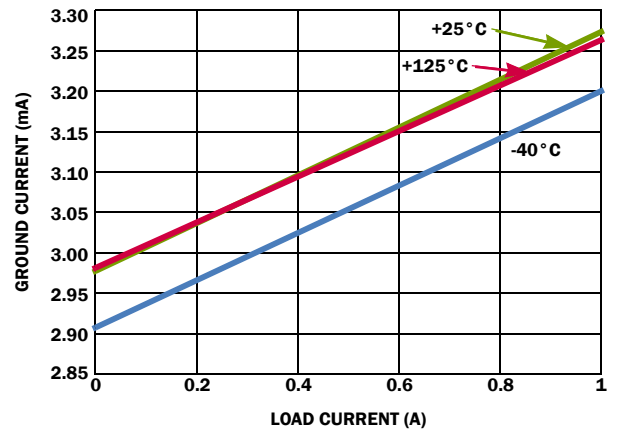


FIGURE 4. GROUND CURRENT vs LOAD CURRENT

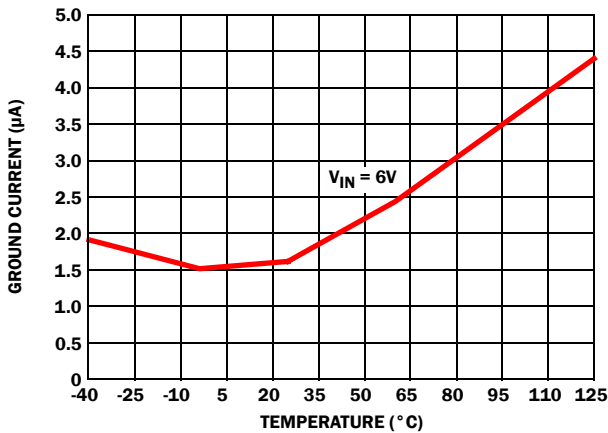


FIGURE 5. SHUTDOWN CURRENT vs TEMPERATURE

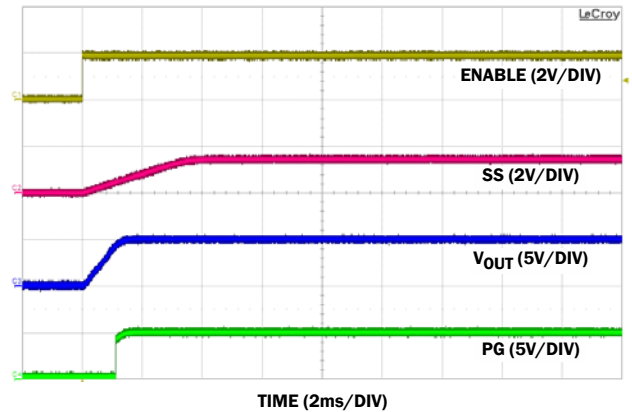


FIGURE 6. ENABLE START-UP

Typical Operating Performance

Unless otherwise noted: $V_{IN} = 5.4V$, $V_{OUT} = 5.0V$, $C_{IN} = C_{OUT} = 10\mu F$, $T_J = +25^\circ C$, $I_L = 0A$. (Continued)

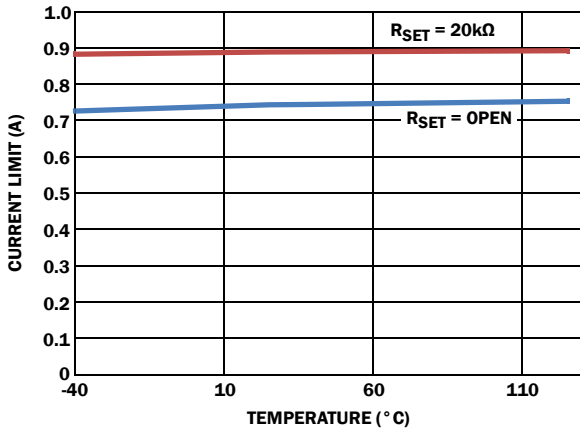


FIGURE 7. CURRENT LIMIT vs TEMPERATURE

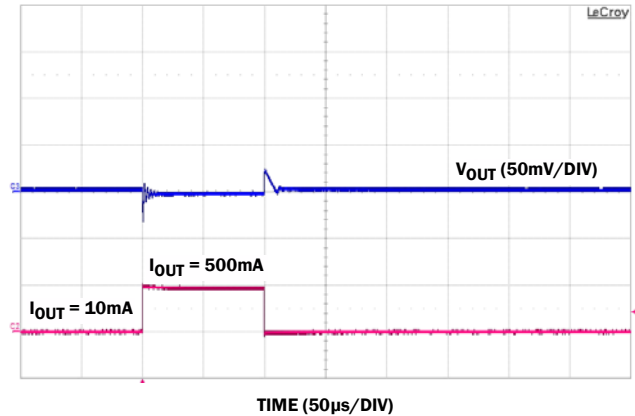


FIGURE 8. LOAD TRANSIENT RESPONSE

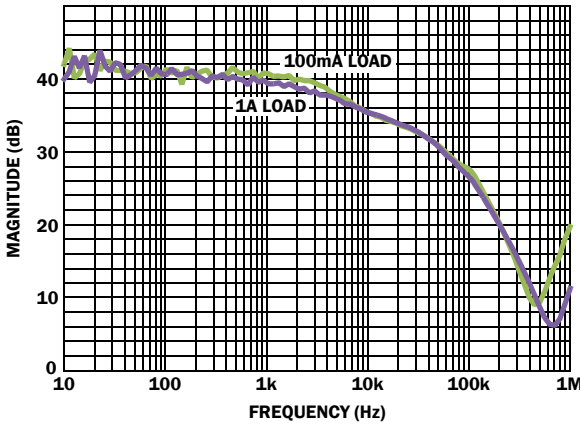


FIGURE 9. PSRR vs LOAD

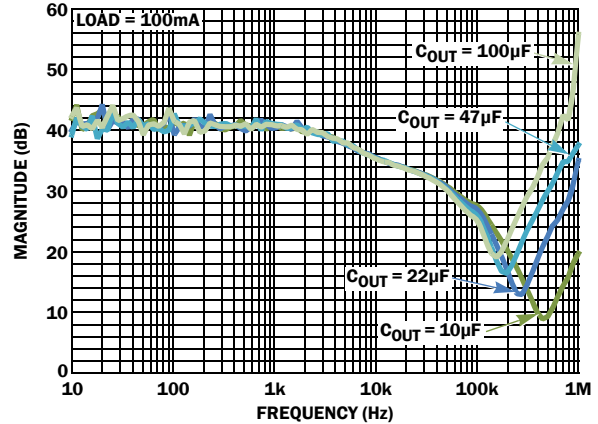


FIGURE 10. PSRR vs C_{OUT}

Functional Description

Input Voltage Requirements

The ISL80121-5 is optimized for 5V output, and can operate from input voltages of 2.2V to 6V. Due to the nature of an LDO, V_{IN} must be some margin higher than V_{OUT} plus dropout at the maximum rated current of the application if active filtering (PSRR) is expected from V_{IN} to V_{OUT} . The generous dropout specification of this family of LDOs allows applications to design for a level of efficiency that can accommodate profiles smaller than the TO220/263.

Programmable Current Limit

The ISL80121-5 protects against overcurrent due to short-circuit and overload conditions applied to the output. When this happens, the LDO performs as a constant current source. If the short-circuit or overload condition is removed, the output returns to normal voltage regulation operation.

The current limit is set at 0.75A by default when the I_{SET} pin is left floating.

This limit can be increased by tying a resistor R_{SET} from the I_{SET} pin to ground. The current limit is determined by R_{SET} as shown in Equation 1:

$$I_{LIMIT} = 0.75 + \frac{2.9}{R_{SET}(k\Omega)} \quad (EQ. 1)$$

Figure 11 shows the relationship between R_{SET} and the current limit when the R_{SET} is tied from I_{SET} pin to GND. Do not short this pin to ground. Increasing the current limit past 1.75A may cause damage to the part and is highly discouraged.

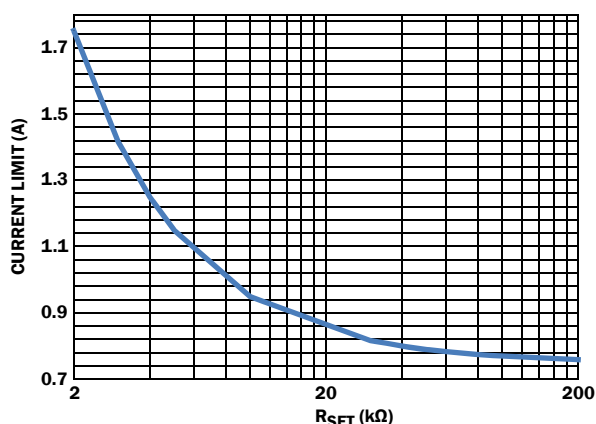


FIGURE 11. INCREASING I_{LIMIT} (R_{SET} TO GND)

The current limit can be decreased from the 0.75A default by tying R_{SET} from the I_{SET} pin to V_{IN}. The current limit is then determined by both R_{SET} and V_{IN} following Equation 2:

$$I_{LIMIT} = 0.75 - \frac{2.9 \times (2 \times V_{IN} - 1)}{R_{SET}(k\Omega)} \quad (EQ. 2)$$

Figure 12 shows the relationship between R_{SET} and the current limit when R_{SET} is tied from the I_{SET} pin to V_{IN} for V_{IN} = 5.4V.

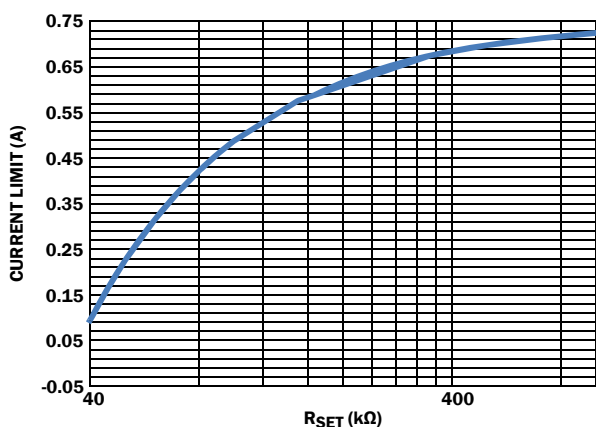


FIGURE 12. DECREASING I_{SET} (R_{SET} TO V_{IN})

Enable Operation

The ENABLE turn-on threshold is typically 800mV with 80mV of hysteresis. An internal pull-up or pull-down resistor to change these values is available upon request. As a result, this pin must not be left floating, and should be tied to V_{IN} if not used. A 1kΩ to 10kΩ pull-up resistor is required for applications that use open collector or open drain outputs to control the ENABLE pin. The ENABLE pin may be connected directly to V_{IN} for applications with outputs that are always on.

Power-Good Operation

PG is a logic output that indicates the status of V_{OUT}, current limit tripping, and V_{IN}. The PG flag is an open-drain NMOS that can sink up to 10mA during a fault condition. The PG pin requires an external pull-up resistor typically connected to the V_{OUT} pin. The PG pin should not be pulled up to a voltage source greater than

V_{IN}. PG goes low when the output voltage drops below 84% of the nominal output voltage, the current limit faults, or the input voltage is too low. PG functions during shutdown, but not during thermal shutdown. For applications not using this feature, connect this pin to ground.

Soft-Start Operation

The soft-start circuit controls the rate at which the output voltage rises up to regulation at power-up or LDO enable. This start-up ramp time can be set by adding an external capacitor from the SS pin to ground. An internal 2μA current source charges up this C_{SS} and the feedback reference voltage is clamped to the voltage across it. The start-up time is set by Equation 3:

$$T_{start} = \frac{(C_{SS} \times 0.5)}{2\mu A} \quad (EQ. 3)$$

Equation 4 determines the C_{SS} required for a specific start-up in-rush current, where V_{OUT} is the output voltage, C_{OUT} is the total capacitance on the output and I_{INRUSH} is the desired in-rush current.

$$C_{SS} = \frac{(V_{OUT} \times C_{OUT} \times 2\mu A)}{I_{INRUSH} \times 0.5V} \quad (EQ. 4)$$

The external capacitor is always discharged to ground at the beginning of start-up or enabling.

External Capacitor Requirements

External capacitors are required for proper operation. Careful attention must be paid to the layout guidelines and selection of capacitor type and value to ensure optimal performance.

OUTPUT CAPACITOR

The ISL80121-5 applies state-of-the-art internal compensation to keep the selection of the output capacitor simple for the customer. Stable operation over full temperature, V_{IN} range, V_{OUT} range and load extremes are guaranteed for all capacitor types and values assuming a minimum of 10μF X5R/X7R is used for local bypass on V_{OUT}. This output capacitor must be connected to the V_{OUT} and GND pins of the LDO with PCB traces no longer than 0.5cm.

There is a growing trend to use very-low ESR multilayer ceramic capacitors (MLCC) because they can support fast load transients and also bypass very high frequency noise from other sources. However, the effective capacitance of MLCCs drops with applied voltage, age, and temperature. X7R and X5R dielectric ceramic capacitors are strongly recommended as they typically maintain a capacitance range within ±20% of nominal voltage over full operating ratings of temperature and voltage.

Additional capacitors of any value in ceramic, POSCAP, alum/tantalum electrolytic types may be placed in parallel to improve PSRR at higher frequencies and/or load transient AC output voltage tolerances.

INPUT CAPACITOR

For proper operation, a minimum capacitance of 10μF X5R/X7R is required at the input. This ceramic input capacitor must be connected to the V_{IN} and GND pins of the LDO with PCB traces no longer than 0.5cm.

Power Dissipation and Thermals

The junction temperature must not exceed the range specified in the “Recommended Operating Conditions” on page 4. The power dissipation can be calculated by using Equation 5:

$$P_D = (V_{IN} - V_{OUT}) \times I_{OUT} + V_{IN} \times I_{GND} \quad (\text{EQ. 5})$$

The maximum allowable junction temperature, $T_{J(\text{MAX})}$ and the maximum expected ambient temperature, $T_{A(\text{MAX})}$ determine the maximum allowable power dissipation, as shown in Equation 6:

$$P_{D(\text{MAX})} = (T_{J(\text{MAX})} - T_A) / \theta_{JA} \quad (\text{EQ. 6})$$

θ_{JA} is the junction-to-ambient thermal resistance.

For safe operation, ensure that the power dissipation P_D , calculated from Equation 5, is less than the maximum allowable power dissipation $P_{D(\text{MAX})}$.

The DFN package uses the copper area on the PCB as a heat-sink. The EPAD of this package must be soldered to the copper plane (GND plane). Figure 13 shows a curve for the θ_{JA} of the DFN package for different copper area sizes.

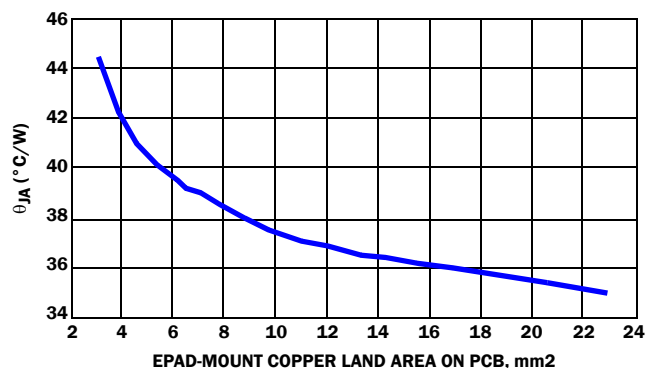


FIGURE 13. 3mmx3mm 10 LD DFN ON 4-LAYER PCB WITH THERMAL VIAS θ_{JA} vs EPAD-MOUNT COPPER LAND AREA ON PCB

Thermal Fault Protection

The power level and the thermal impedance of the package (+48 °C/W for DFN) determine when the junction temperature exceeds the thermal shutdown temperature. In the event that the die temperature exceeds around +160 °C, the output of the LDO will shut down until the die temperature cools down to about +130 °C.

General PowerPAD Design Considerations

Figure 14 shows the recommended use of vias on the thermal pad to remove heat from the IC. This typical array populates the thermal pad footprint with vias spaced three times the radius distance from the center of each via. Small via size is advisable, but not to the extent that solder reflow becomes difficult.

All vias should be connected to the pad potential, with low thermal resistance for efficient heat transfer. Complete connection of the plated-through hole to each plane is important. It is not recommended to use “thermal relief” patterns to connect the vias.

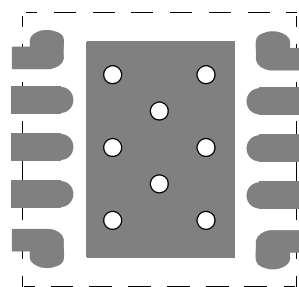


FIGURE 14. PCB VIA PATTERN

For additional products, see www.intersil.com/product_tree

Intersil products are manufactured, assembled and tested utilizing ISO9000 quality systems as noted in the quality certifications found at www.intersil.com/design/quality

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Revision History

The revision history provided is for informational purposes only and is believed to be accurate, but not warranted. Please go to web to make sure you have the latest revision.

DATE	REVISION	CHANGE
4/22/11	FN7713.3	In Figure 8 on page 7, corrected label from "V _{OUT} (50V/DIV)." to "V _{OUT} (50mV/DIV)." In "DC Output Voltage Accuracy" on page 4, corrected the MAX value from -1.8 to +1.8.
2/1/11	FN7713.2	<ol style="list-style-type: none"> page 1, paragraph 2, "The programmable current limiting improves system reliability of applications" changed to "The programmable current limiting improves system reliability of end applications." page 1, Features, "Programmable Soft-starting" changed to "Programmable Soft-Start" Made subbing consistent throughout document. page 3, EPAD Description "directly to GND plane is optional." Changed to "directly to GND plane is required for thermal considerations. See "Power Dissipation and Thermals" on page 9 for more details." page 5, Removed Notes in Electrical Spec Table, which read : "Minimum capacitor of 10µF X5R/X7R on VIN and VOUT required for stability." and "If the current limit for in-rush current is acceptable in application, do not use this feature. Used only when large bulk capacitance required on VOUT for application." page 5, Electrical Specifications, PG Pin Characteristics, Vout PG Flag Threshold <ol style="list-style-type: none"> Typical "85" changed to "84" %Vout page 9, after Thermal Fault Protection section <ol style="list-style-type: none"> Added "General PowerPAD Design Considerations" section with Figure 14. All PGOOD changed to PG throughout.
1/28/11		Changed Theta Ja from 51C/W to 48C/W.
1/25/11		<ol style="list-style-type: none"> page 1, Features <ol style="list-style-type: none"> "200µVrms Output Noise" changed to "210 µVrms Output Noise" page 1, Typical Applications, right side figure <ol style="list-style-type: none"> Resize "VOUT" (pin2) and "SENSE" (pin3) page 8, Equation 4 <ol style="list-style-type: none"> Extra parenthesis ")" removed
1/21/11		<p>page 1 Before Features added Table of Key Differences.</p> <p>page 2 Block Diagram - Removed "ADJ Voltage Version" and left the "Sense" Connection.</p> <p>page 3 Pin Number 8, description, 2nd sentence: "Current limit is 0.75mA..." changed to "Current limit is 0.75A..."</p> <p>page 4 Electrical Specifications, AC Characteristics, Input Supply Ripple Rejection Test conditions and Typical values changed from "f = 1kHz, ILOAD = 1A, f = 120Hz, ILOAD = 1A" TO "f = 1kHz, ILOAD = 1A, f = 1kHz, ILOAD = 100mA"</p> <p>page 6, Figure 3 - X-axis label changed from "Output Current (mA)" to "Output Current (A)"</p> <p>page 8, Figure 12 - a. Figure label change. "IN" in "VIN" was subscripted.</p>
12/6/10	FN7713.1	<ol style="list-style-type: none"> In "Block Diagram" on page 2: <ol style="list-style-type: none"> Added "ADJ adjustable voltage version" Pin. Added "fixed voltage version" to "SENSE" pin On page 4: "Ground Pin Current" Test Conditions <ol style="list-style-type: none"> Replaced "V_{OUT}+0.4V" with "2.2V" on both lines
12/2/10	FN7713.0	Initial Release.

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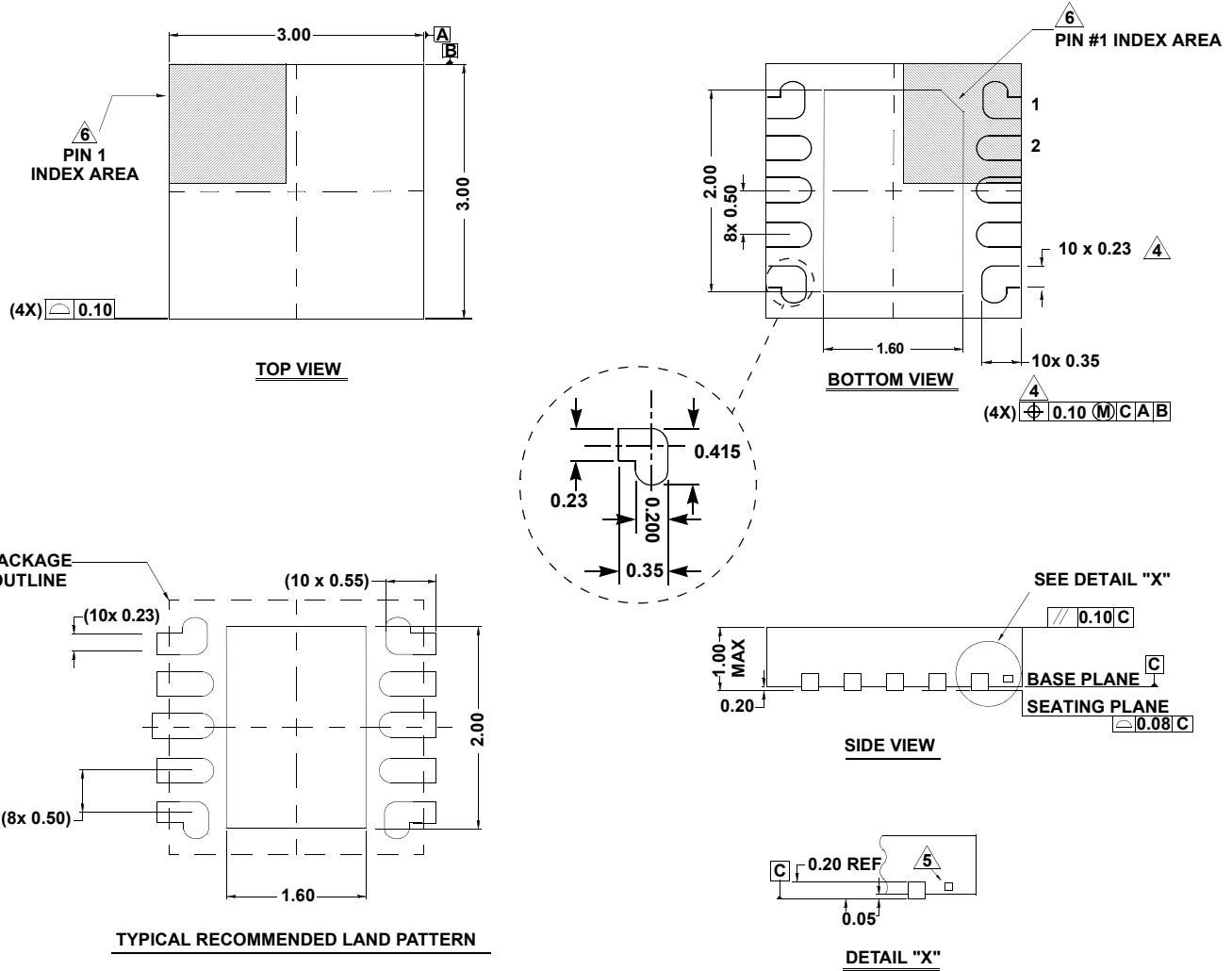
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Package Outline Drawing

L10.3x3

10 LEAD DUAL FLAT PACKAGE (DFN)

Rev 6, 09/09



NOTES:

- Dimensions are in millimeters.
Dimensions in () for Reference Only.
- Dimensioning and tolerancing conform to AMSE Y14.5m-1994.
- Unless otherwise specified, tolerance : Decimal ± 0.05
- Lead width applies to the metallized terminal and is measured between 0.18mm and 0.30mm from the terminal tip.
- Tiebar shown (if present) is a non-functional feature.
- The configuration of the pin #1 identifier is optional, but must be located within the zone indicated. The pin #1 identifier may be either a mold or mark feature.